Implementation of waveform compression for ICECUBE

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LBNL

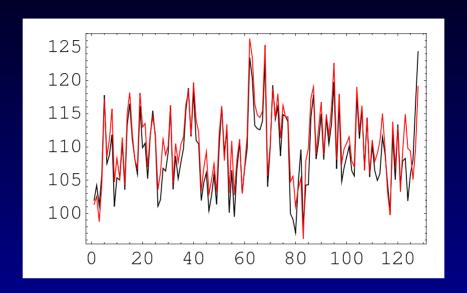
Constraints on compression

- Bandwidth: 1 Mbaud per 2 DOMs → Karl-Heinz Sulanke
- Computing resources:
 - CPU: 40 MHz ARM (16.8 MHz in String-18)
 - FPGA: \sim 16,000 Logic elements (\sim 3000 in String-18, no room left for compression)
- Noise rate: 500 Hz (required; 1kHz in String-18)
- Correlation of noise:
 - ⇒ feature extract SPEs and send MPEs uncompressed doesn't work.

Principle idea

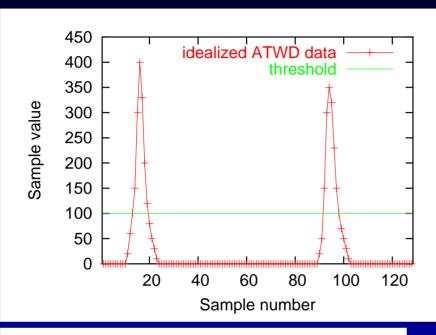
- Waveform of 1 ATWD channel or fADC similar to a facsimile scan line.
- Waveform corresponds to gray-scale representation of FAX line.
- Original FAX encoding: run-length encoding followed by Huffman encoding.
- Group 3 CCITT facsimile standard: fixed, immutable, Huffman code, optimized for a set of eight standard documents.
- Do Huffman encoding of waveforms with static table (to be generated as a calibration task).
- Further simplify this to "Huffman-lite"

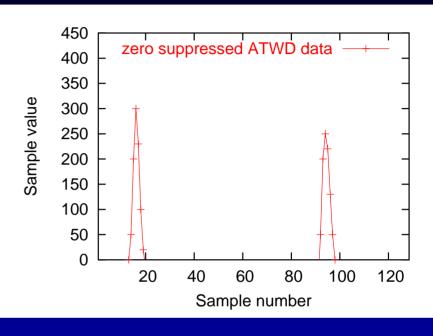
Step I: Pedestal etc

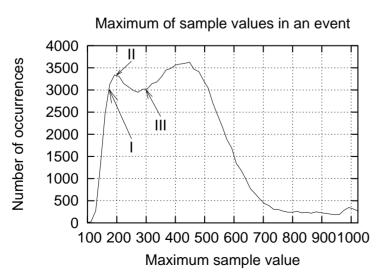


- Pedestal investigated by David Seckel: not perfectly constant in String-18, integrated IceCube DOM should deliver constant fingerprint (ATWD tester, STF, TestDAQ)
- Ringing should not occur in IceCube DOMs!
- Baseline: Low frequency noise on different ground levels.
- Pick ATWD channel with highest gain that doesn't saturate.

Step II: Zero suppression

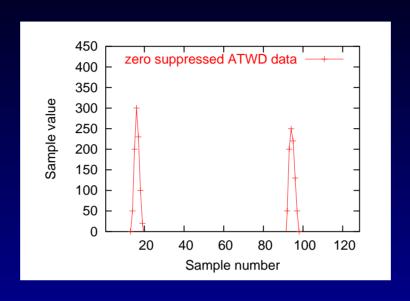






Typical setting of threshold of discriminator and zero suppression: 25% - 33% of average SPE pulse hight over baseline.

Step III: Run-length encoding



0...0 50 200 300 230 100 20 0...0 50 200 250 220 130 50 0...0: **128 Bytes**



12 0, 0 50, 0 200, 0 300, 0 230, 0 100, 0 20, 71 0, 0 50, 0 200, 0 250, 0 220, 0 130, 0 50, 30 0: **30 Bytes**

Meaning of pair here: Number of consecutive *repetitions*, Value. Standard RL: Number of consecutive *occurencies*, Value.

tweak of standard $\Rightarrow \sim 50\%$ of numbers are zeros (less entropy).

Step IV: Huffman-lite

Huffman encoding:

- Minimal variable-length "character" encoding based on the frequency of each "character".
- more frequent "characters" are encoded with few bits, and rare "characters" are encoded with many bits.

Huffman-lite:

- $\sim 50\%$ of numbers ("characters") are zeros.
- Minimize bits used for zeros, don't work on finite values.
- Convert "0" (8 bits) \rightarrow '0' (1 bit) Convert "N" (8 bits) \rightarrow '1', "N" (9 bits) $12\ 0, 0\ 50, \ldots \rightarrow$ '1000011000100110010 \ldots '

Compression efficiency

[Bytes/event]	SIRING-18			ICECUBE (estimate)		
Method	ATWD	fADC	header	ATWD	fADC	header
RAW	128	128	6	128	128	6
run-length	19	6	6	11 ^a +3 ^b	6	6
"gzip –fast"	15		6			6
'bzip2 -9''	10		6			6
'Huffman-lite''	15		6	12		6

Total: "run-length + Huffman lite" 21

18

ICECLIBE (actimata)

- STRING-18: 1 kHz PMT noise, 1 DOM/cable \rightarrow 21 kB/s
- ICECUBE (estimate): 0.5 kHz, $2 \text{ DOM/cable} \rightarrow 18 \text{ kB/s}$
- Available bandwidth (IceCube): 100 kB/s/cable

^a 11: Half the sampling speed of ATWD.

b 3: Estimate for longer ATWD time window in ICECUBE.

Computing requirements

- Intel Pentium-4:
 - \sim 10,000 CPU clock cycles per event.
 - Pedestal subtraction, zero supp. (Step I&II): $\sim 50\%$.
 - Run-lenght encoding (Step III): $\sim 30\%$.
 - Huffman encoding (Step IV): $\sim 20\%$.
- ARM on IceCube MB:
 - \sim 20,000 CPU clock cycles per event.
- GTP & TS confident: Whole algorithm fits in FPGA.
- Certainly, biggest chunk for CPU (Step I) fits.

IceCube MB CPU: 40 MHz (can be increased); 500 Hz noise ⇒ CPU load due to compression < 15%.

Induced losses

- Step I (pedestal pattern etc): lossless.
- Step II (zero suppression): lossy!
 - Concept in particle physics for half a century.
 - Obvious way to distinguish electronic noise from PMT.
 - Equivalent to discriminator effect if threshold is set the same "value": we already accepted to lose these pulses below threshold.
- Step III (Run-length): lossless.
- Step IV (Huffman-lite): lossless.
 - \Rightarrow I don't understand the excitement about the losses!

Room for improvements

- Full Huffman encoding (\sim factor 1.5)
- Reduce resolution of sample value (→ charge):
 PDD: 5%; 10 bits: 0.1%, 8 bits: 0.4%, 4 bits: 6%
- Reduce time resolution:
 PDD: 5 ns; ATWD sampling speed: 3.3 ns; going to 6.7 ns and fitting the pulse shape
 → still meet the requirement.
- Going to assembler (instead of C)
 → reduce CPU cycles.

Conclusion

- Algorithm is simple.
- Bandwidth: Factor of 5 safety margin!
- Several parameters left to improve further.
- Computing resources (CPU, FPGA) sufficient.
- Induced losses: Already accepted!
- Some implementation details need to be sorted out (header alignment, where to decompress at surface ...)

 - Compression is on sound basis
 - My opinion: "Local coincidence" obsolete if glass/noise and cable meet requirements.